

A Noise-Adaptive Algorithm for First-Order Reed-Muller Decoding

Jon Feldman

jonfeld@mit.edu

Matteo Frigo

athena@vanu.com

Ibrahim Abou-Faycal

iaboufay@mit.edu

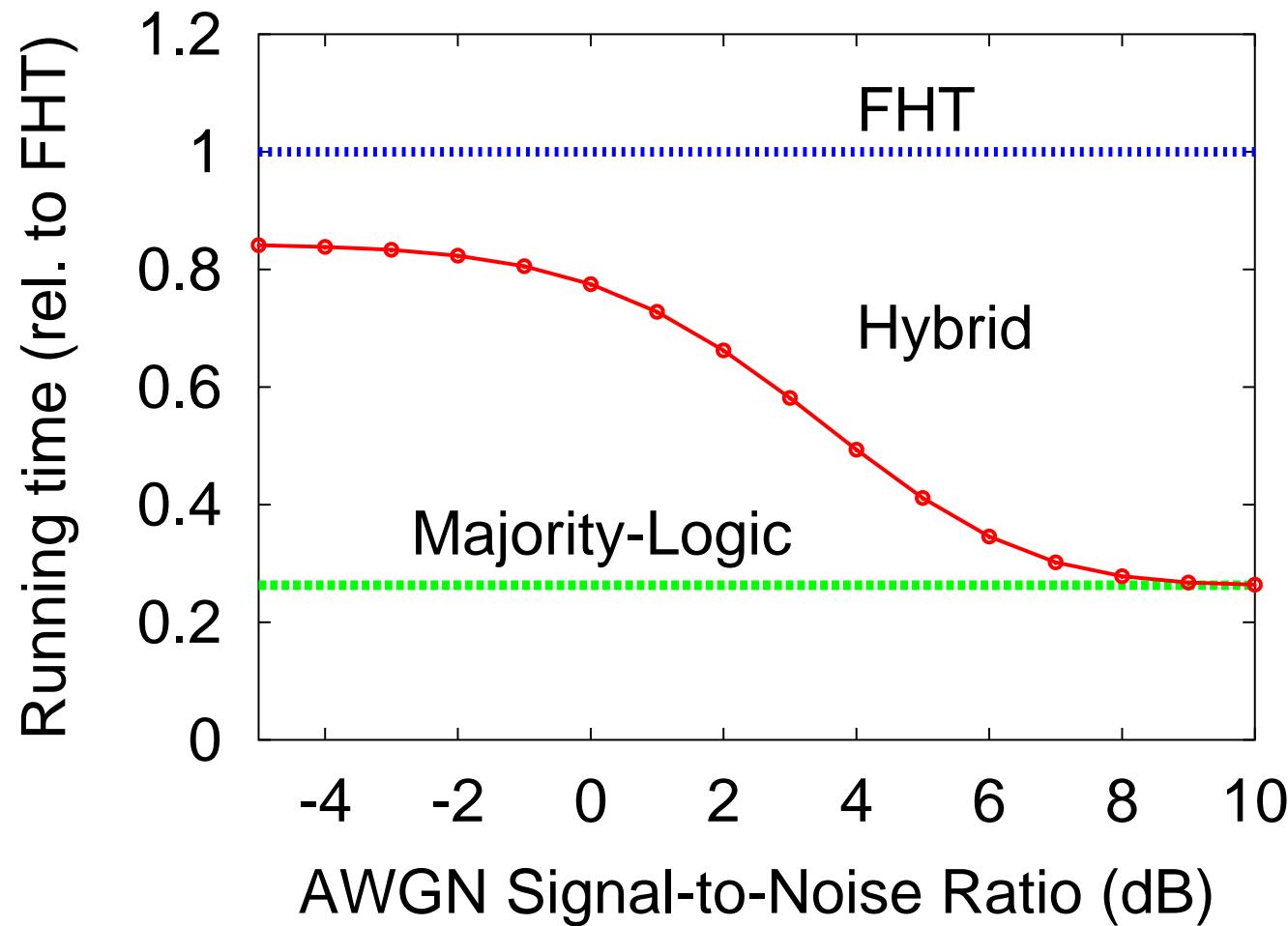
M.I.T.

Vanu, Inc.

CCK demodulation

- Demodulation: bottleneck in software 802.11b implementation.
- Standard optimal demodulator based on Fast Walsh-Hadamard transform (FHT);
 - Software radios cannot take advantage of parallelism.
- Majority-logic demodulators [Reed '54, Massey '63] efficient but suboptimal.
- Our *Hybrid* algorithm:
 - almost as fast as majority-logic;
 - “almost as optimal” as FHT.

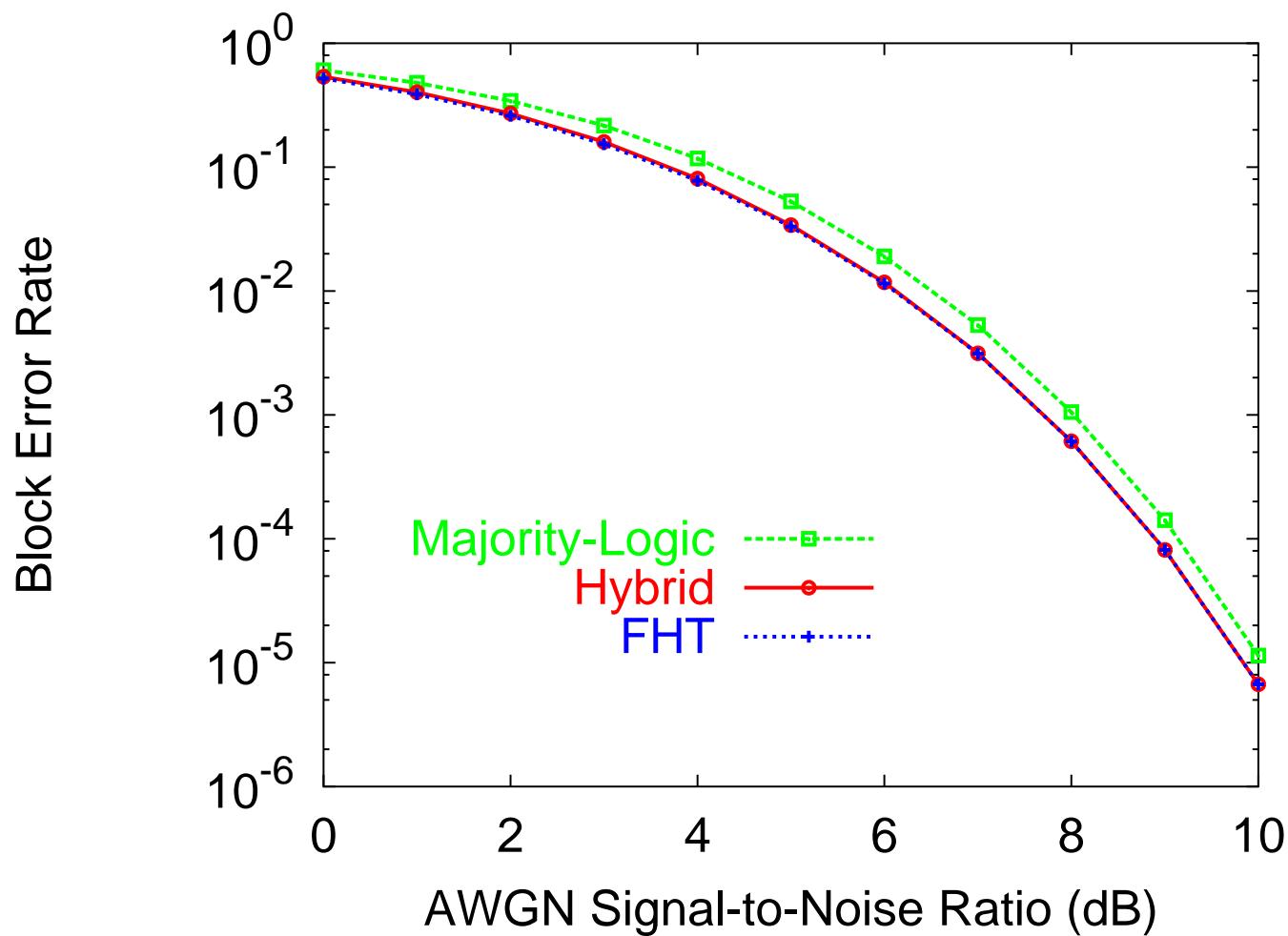
Running Time Comparison



- Running time (implicitly) SNR-dependent
 - OK for software radios.

Performance Comparison

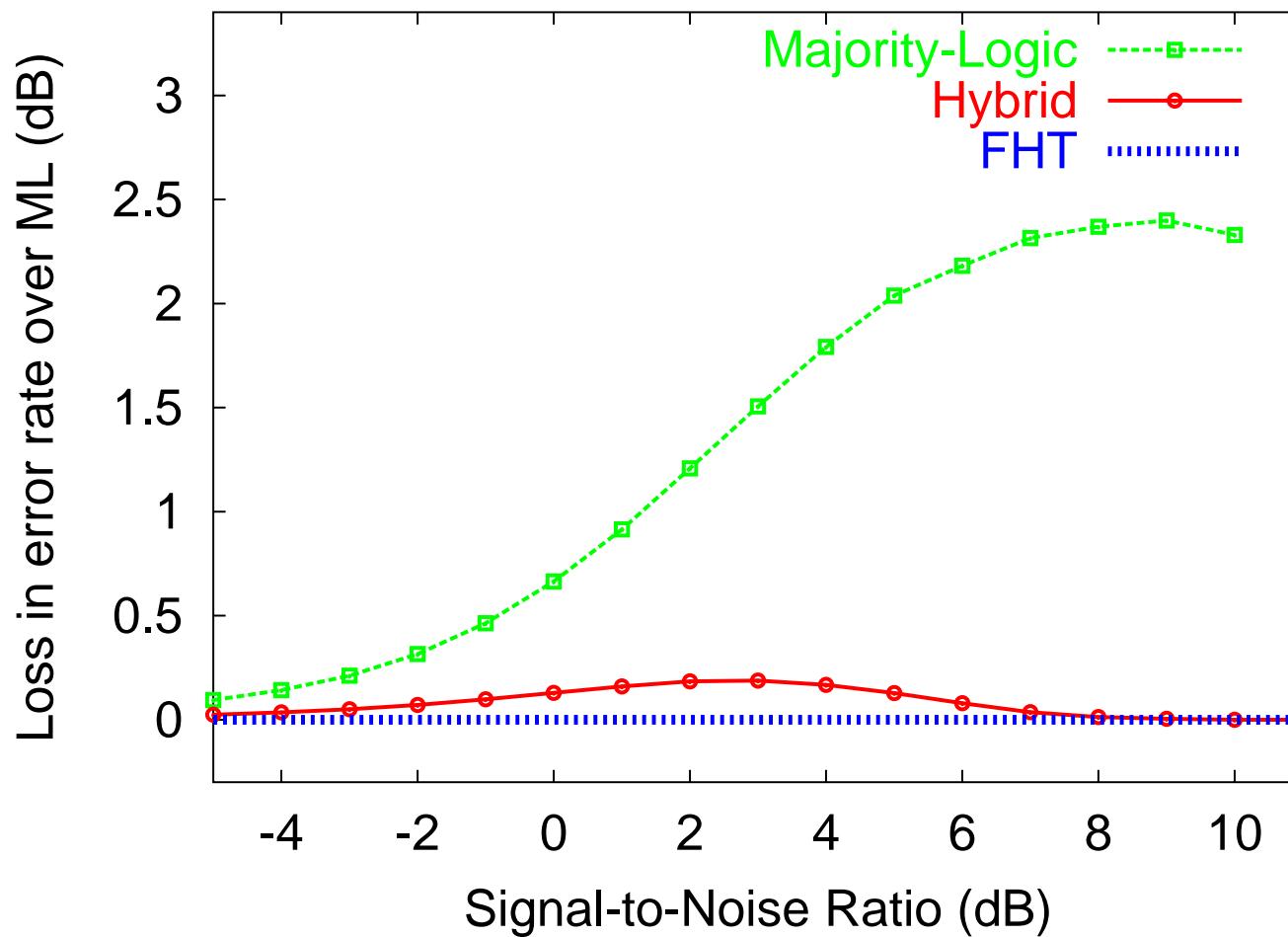
CCK Demodulation



- Hybrid algorithm very close to optimal FHT.

Performance Comparison (closer look)

CCK Demodulation



- Negligible loss of performance (≤ 0.2 dB).

Outline

- CCK modulation / demodulation.
- Majority logic decoding.
- The hybrid algorithm.
- Generalization to first-order Reed-Muller (FORM) codes:
 - H_e : Error rate of Hybrid algorithm.
 - O_e : Error rate of ML decoder (FHT).

$$H_e \leq O_e + \exp(-\Omega(n)).$$

CCK modulation

- **Info:** 4 “complex bits:”

$$\phi = (\phi_0, \phi_1, \phi_2, \phi_3) \quad (\phi_i \in Q, \quad Q = \{1, i, -1, -i\})$$

- **Transmit:** $x(\phi) = (x_0, \dots, x_7)$, where*:

$$\begin{array}{ll} x_0 = \phi_3 & x_4 = \phi_3 \quad \phi_2 \\ x_1 = \phi_3 & \phi_0 \qquad \qquad x_5 = \phi_3 \quad \phi_2 \quad \phi_0 \\ x_2 = \phi_3 \quad \phi_1 & x_6 = \phi_3 \quad \phi_2 \quad \phi_1 \\ x_3 = \phi_3 \quad \phi_1 \quad \phi_0 & x_7 = \phi_3 \quad \phi_2 \quad \phi_1 \quad \phi_0 \end{array}$$

- **Receive:** (y_0, \dots, y_7) , $y_i = x_i + N_i(0, \sigma^2)$
- (* In real system, x_1 and x_4 negated.)

CCK demodulators

- Maximum-Likelihood decoding: find ϕ^{\max} where

$$\phi^{\max} = \max_{\phi \in Q^4} |\textcolor{red}{x}(\phi) \cdot \textcolor{magenta}{y}|, \quad (Q = \{1, i, -1, -i\}).$$

- Can be computed via Fast Hadamard Transform (FHT).
- FHT not fast enough for software radio.

- Majority-Logic [Reed '54, Massey '63] decoding:
 - Extract “votes” for each information symbol.
 - Tally votes, majority rules for each symbol.
 - Use for CCK: [van Nee '96, Paterson/Jones '98].

Majority-Logic Decoding for CCK

$$x_0 = \phi_3$$

$$x_1 = \phi_3 \quad \phi_0$$

$$x_2 = \phi_3 \quad \phi_1$$

$$x_3 = \phi_3 \quad \phi_1 \quad \phi_0$$

$$x_4 = \phi_3 \quad \phi_2$$

$$x_5 = \phi_3 \quad \phi_2 \quad \phi_0$$

$$x_6 = \phi_3 \quad \phi_2 \quad \phi_1$$

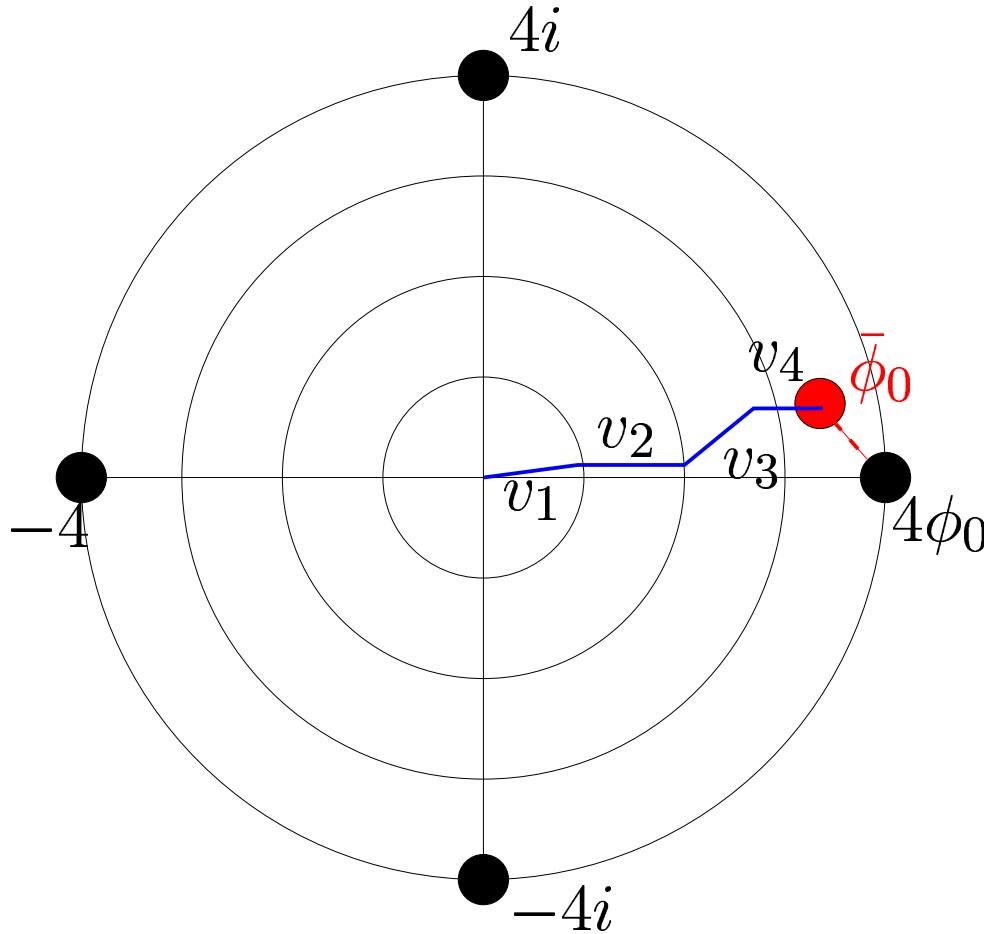
$$x_7 = \phi_3 \quad \phi_2 \quad \phi_1 \quad \phi_0$$

- Example: $x_3 x_2^* = (\phi_3 \phi_1 \phi_0)(\phi_3 \phi_1)^* = \phi_3 \phi_1 \phi_0 \phi_3^* \phi_1^* = \phi_0$
- This makes $y_3 y_2^*$ a “vote” for ϕ_0 :

$$\begin{aligned} E[y_3 y_2^*] &= E[(x_3 + N_3)(x_2 + N_2)^*] \\ &= E[(x_3 + N_3)] E[(x_2^* + N_2^*)] \\ &= x_3 x_2^* \\ &= \phi_0 \end{aligned}$$

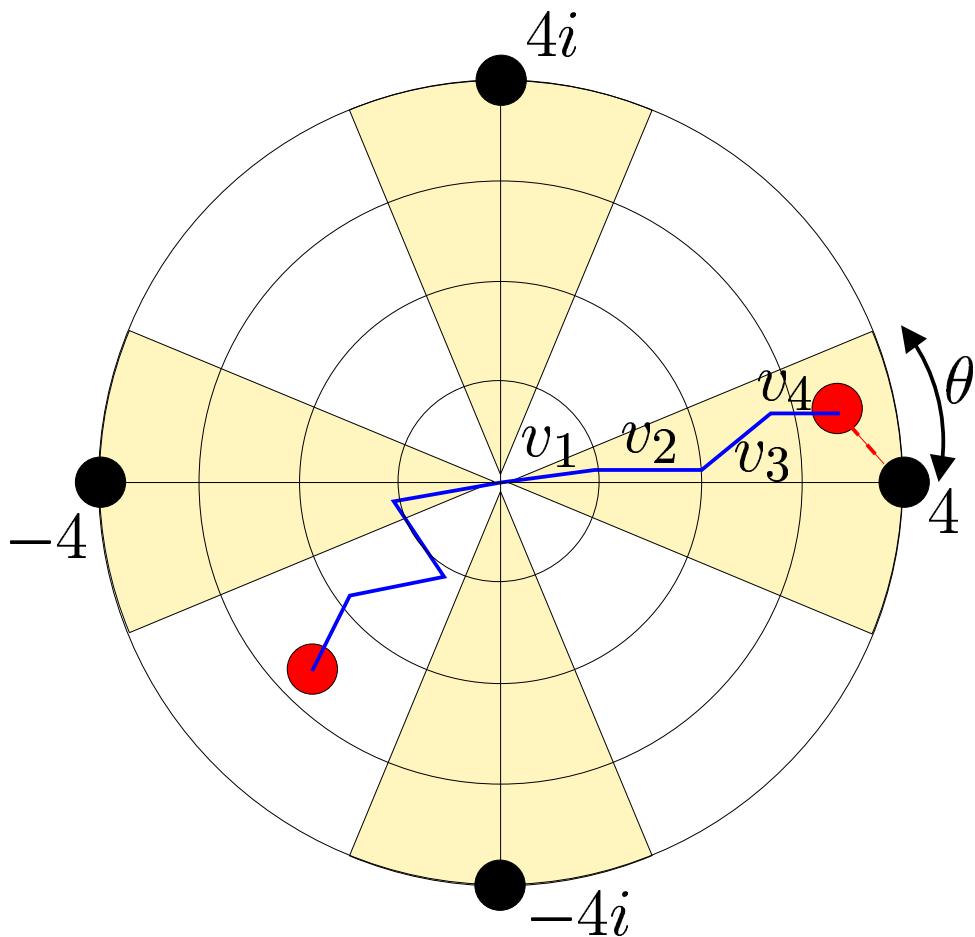
- Four independent votes for ϕ_0 : $y_1 y_0^*$, $y_3 y_2^*$, $y_5 y_4^*$, $y_7 y_6^*$

Tallying “Soft” Votes



- Suppose $\phi_0 = 1$.
- Four votes:
$$v_1 = y_1 y_0^* \quad v_2 = y_3 y_2^* \\ v_3 = y_5 y_4^* \quad v_4 = y_7 y_6^*$$
- “Estimate” $\bar{\phi}_0$:
$$\begin{aligned} \bar{\phi}_0 &= \sum_{i=1}^4 v_i \\ &\approx 4\phi_0 \\ &= 4 \end{aligned}$$
- Set ϕ_0 to “closest” point in $\{4, 4i, -4, -4i\}$.

The Hybrid Algorithm



- Set “threshold angle” θ .
 - $\theta = \tan^{-1}(2/3)$
- Compute est. $\bar{\phi}_0, \bar{\phi}_1, \bar{\phi}_2$.
- Find closest $\phi_i \in Q$ for each estimate:
$$\phi_i = \arg \min_{\phi \in Q} |\angle(\phi) - \angle(\bar{\phi}_i)|.$$
- If $|\angle(\bar{\phi}_i) - \angle(\phi_i)| > \theta$ for any $i \in \{0, 1, 2\}$, run FHT.
- Otherwise, compute ϕ_3 from ϕ_0, ϕ_1, ϕ_2 .

General FORM Codes

- Definition of $FORM_q(k, p)$:
 - Information word $c \in \mathbb{Z}_q^k$.
 - Polynomial $P(x) = c^T x$, where $x \in \{0, \dots, p-1\}^k$ for some $p \leq q$.
 - Codeword: Evaluate $P(x) \bmod q$ for all possible values of x . Code length $n = p^k$.
- Classic Reed-Muller codes: $p = 2$.
- Hadamard Codes: $FORM_2(k, 2)$.
- CCK: isomorphic to $FORM_4(3, 2)$.
- Generalized version of hybrid algorithm works for any FORM code.

Coding Theorem for AWGN Channel

- H_e : Error rate of Hybrid algorithm.
- O_e : Error rate of ML decoder (FHT).
- Theorem: For all $0 < \alpha < 1$, $0 < t < 1$,

$$H_e \leq O_e + \exp(-A_1 n) + \exp(-A_2 n).$$

$$\begin{aligned} A_1 &= \frac{(1 - \alpha)^2 \sin^2(2\pi/q - \theta)}{8\sigma^2} \\ A_2 &= \frac{1}{2} \left(\frac{t\alpha \sin(2\pi/q - \theta)}{\sigma^2} - \ln \left(\frac{t \arccos(-t)}{(1 - t^2)^{3/2}} + \frac{1}{1 - t^2} \right) \right) \end{aligned}$$

- Example: $q = 4$ (QPSK), $\theta = \tan^{-1}(2/3)$, $\text{SNR} > 4 \text{ dB}$,

$$H_e \leq O_e + 2^{1-n/10}.$$

Conclusion

- Hybrid algorithm for CCK:
 - Provides “near-optimal” decoding,
 - Runs at a fraction of the running time,
 - Allows software implementation of 802.11b.
- For any FORM code:

$$H_e \leq O_e + \exp(-\Omega(n)).$$